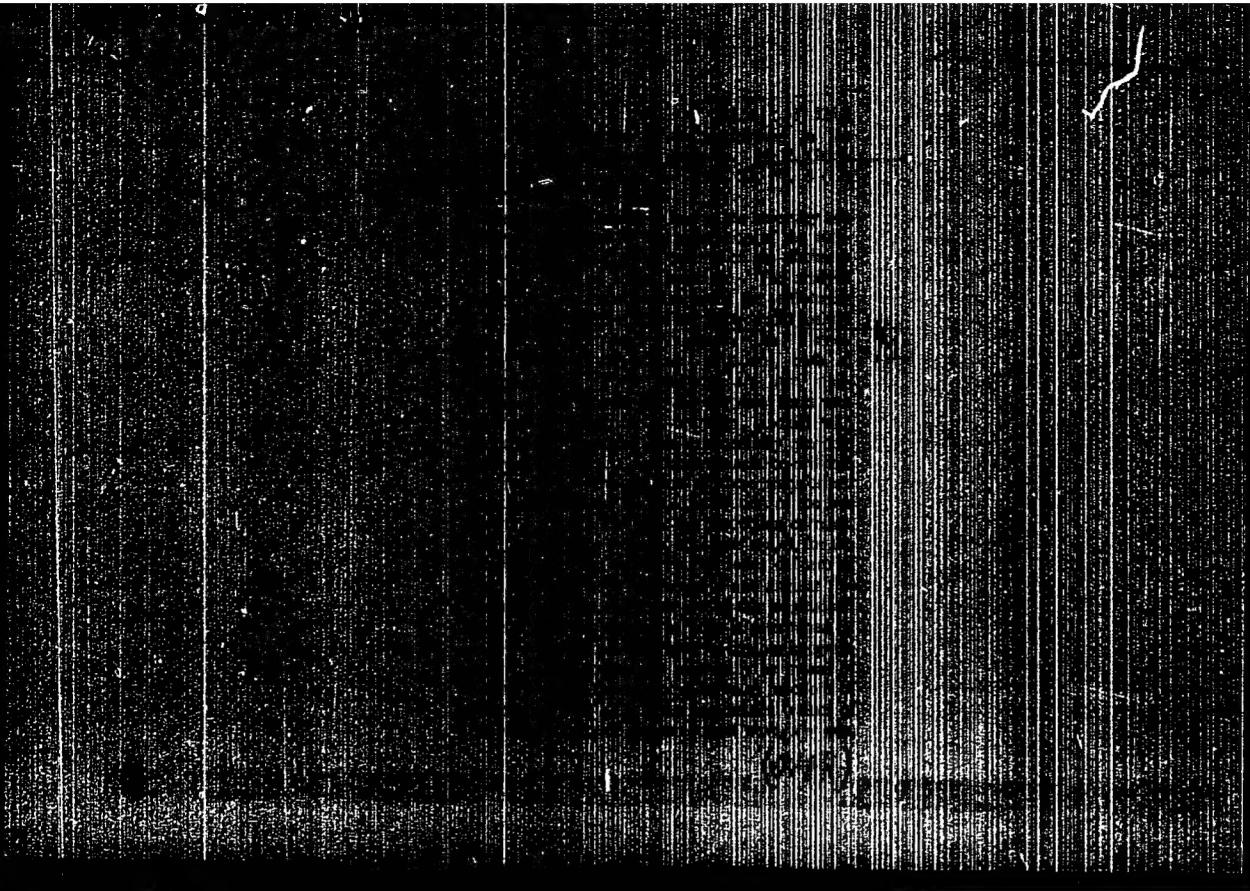


"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2

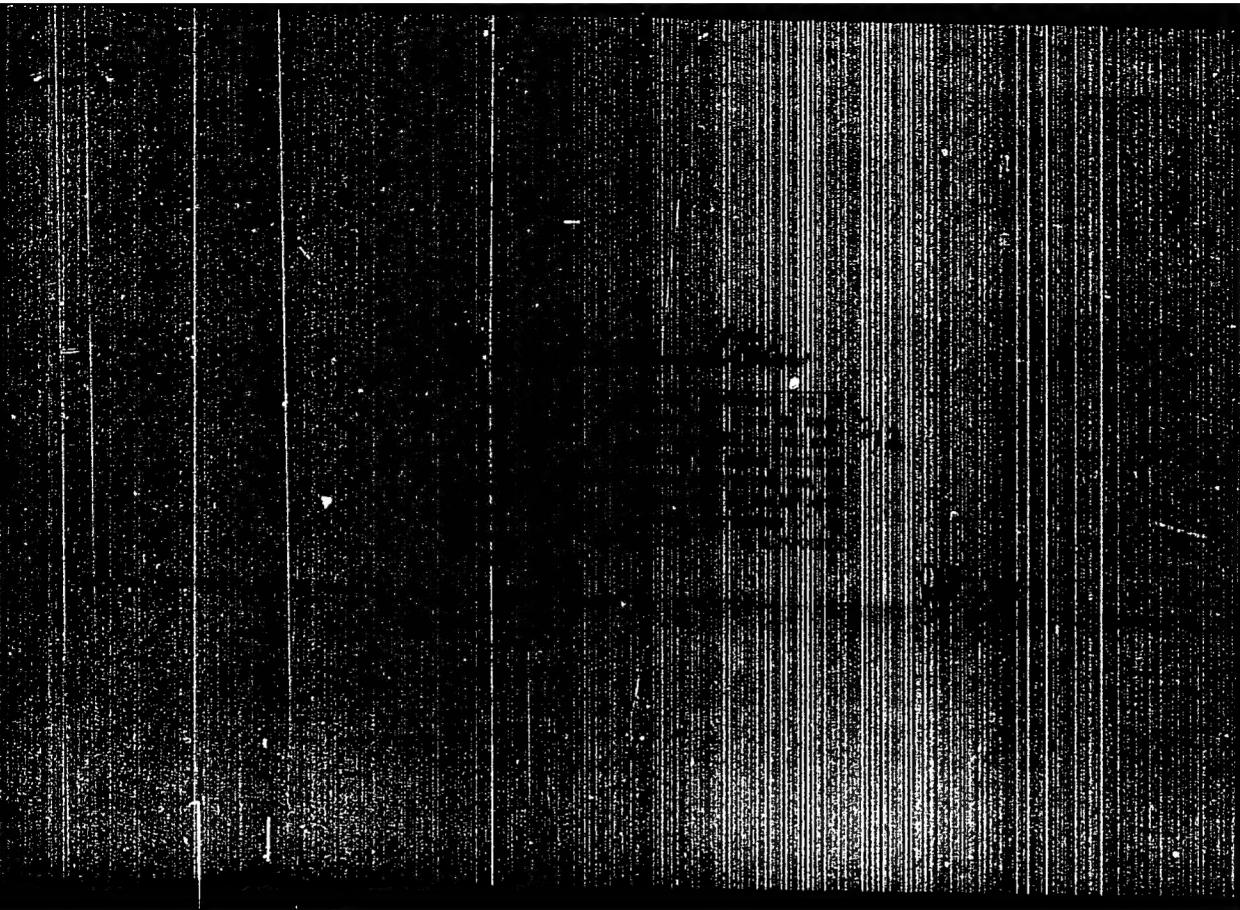


APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2

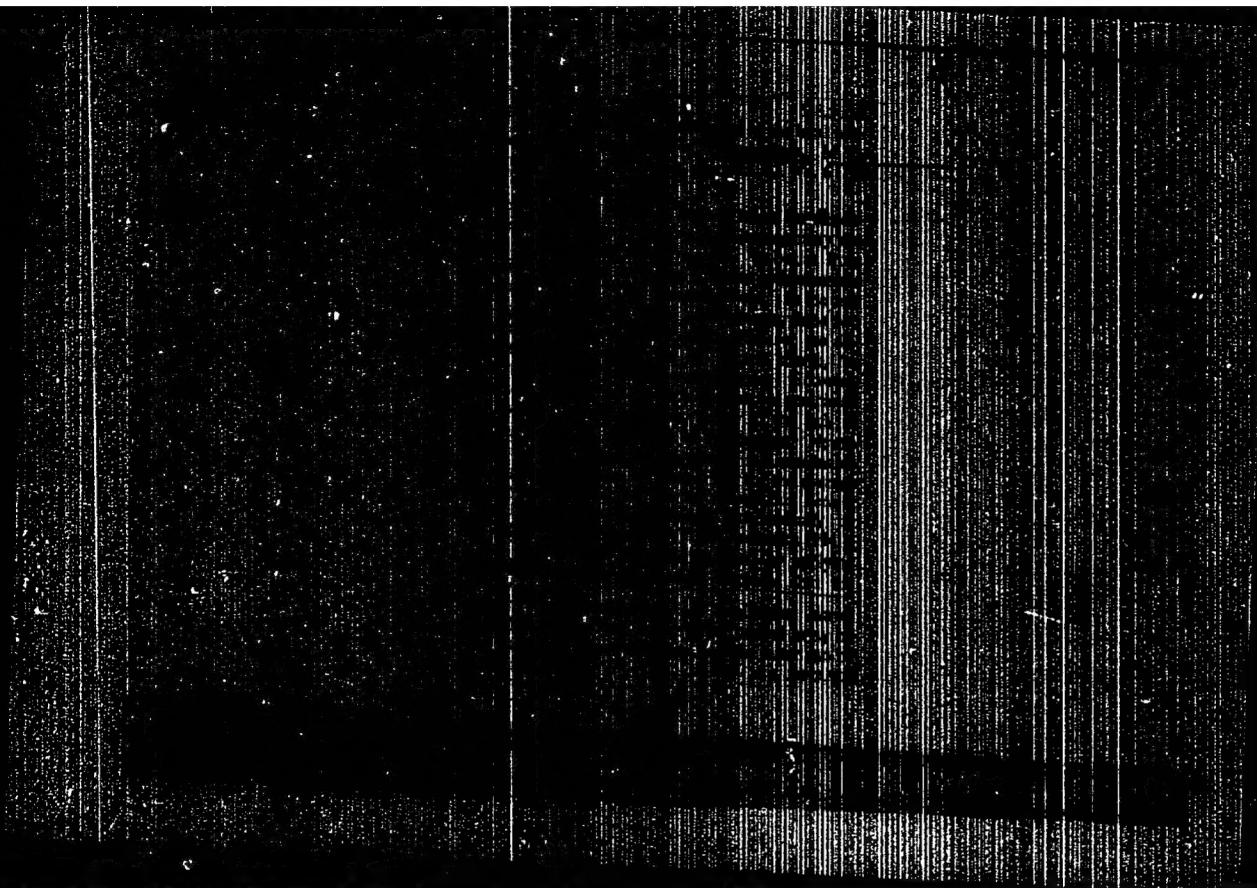


APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

GERTSENSHTEYN, M.Ye.
POTEMKIN, V.V.; GERTSENSHTEYN, M.Ye.

G.V.Gordeev's strata theory. Zhur.ekspl. i teor.fiz. 24 no.5:610-612
Mv '53. (MLRA 7:10)
(Nuclear physics)

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2

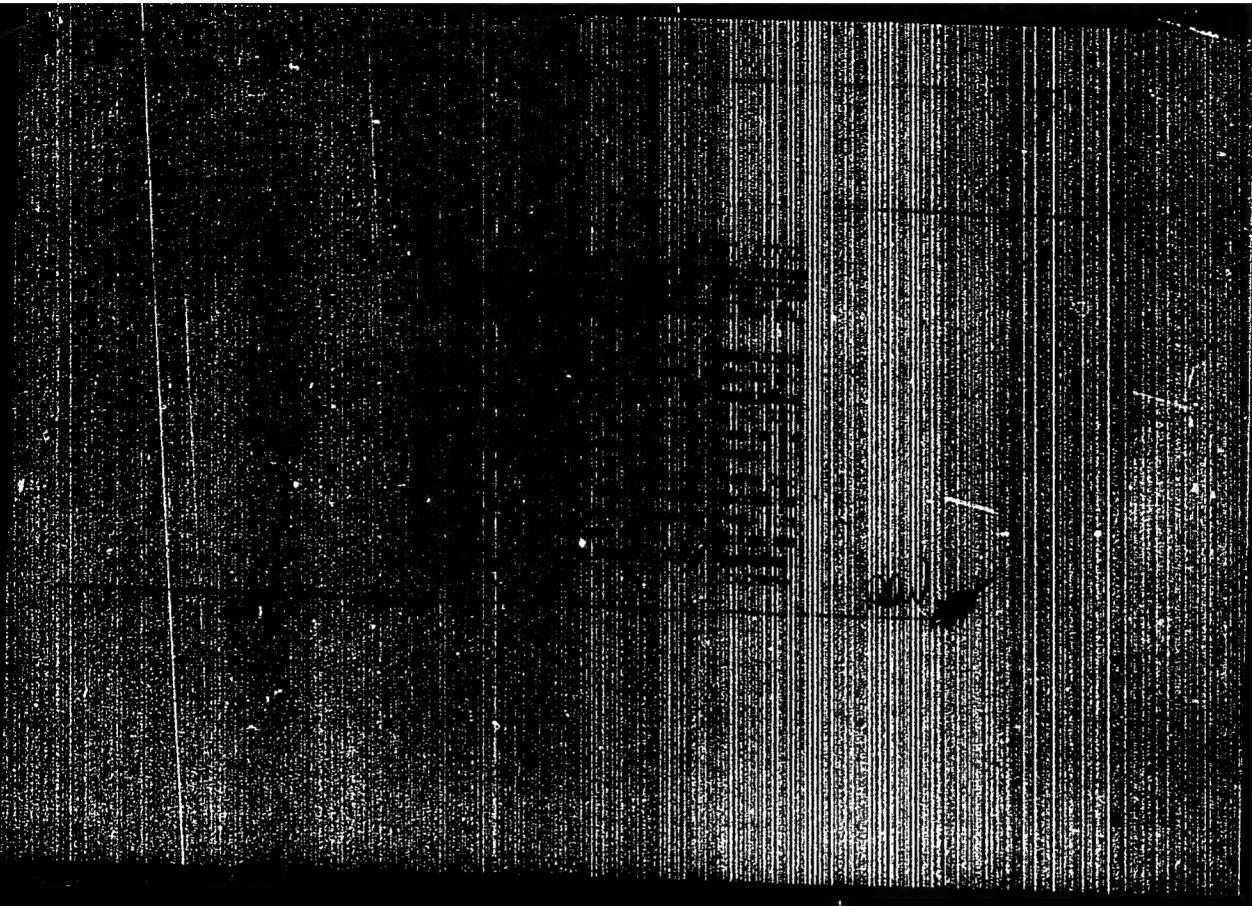


APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

GERTSENSTEYN, M. Ye.
USSR/Physics - Self-excited oscillations

FD 405

Card 1/1

Author : Gertsenshteyn, M. Ye.
Title : Self-excited oscillations in gaseous discharge at high pressures
Periodical : Zh. eksp. i teor. fiz. 36, 57-63, Jan 1954
Abstract : Treats the interaction of sound waves and electron waves in gas-discharge plasma. Demonstrates the possibility of self-excited oscillations for a definite interval of frequencies. Thanks V. V. Potemkin for his judgment of the physical results. Fourteen references, including K. F. Teodorchik, *avtokolejotai'nyye sistemy* (Self-excited oscillator systems), State Technical-Theoretical Literature Press, 1952.
Institution : Moscow State University
Submitted : November 1, 1954.

GERTSEN SHTEYN, M. Ye.
ISSP/ Physics - Electrodynamics

FD-15

Card 1/1 : Pub 146-3/1a

Author : Gertsen Shteyn, M. Ye.

Title : Energy current in spatial dispersing media

Periodical : Zhur. ekspl. i teor. fiz., 26, 666-680, Jun. 1954.

Abstract : S. M. Rytov's results (ibid. 17, 110 (1947)) are generalized to the case of spatial dispersion when the partial derivative is not zero. It is shown that in this case the velocity of energy propagation coincides with the group velocity. + references. Indebted to V. V. Potemkin.

Institution : --

Submitted : October 15, 1952

GERTSENSHTEYN, M. YE
USSR/Physics - Plasma

Card 1/1 Pub. 146-8/21

FD-795

Author : Gertsenshteyn, M. Ye.
Title : Dielectric permeability of plasma located in a stationary magnetic field
Periodical : Zhur. eksp. i teor. fiz., 27, 180-186, Aug 1954
Abstract : The tensor of the complex dielectric permeability of an electron gas is
computed taking into account the thermal motion of electrons. Indebted
to V. V. Potemkin. Sixteen references, including 3 foreign
Institution : Central Scientific Research Institute of Radio Measurements
Submitted : October 15, 1953

GERTSENSHTEIN, M. V.

✓ Low-frequency oscillations in the positive column of a glow discharge. M. E. Gertsenshtain and V. V. Potemkin (Moscow State Univ.). *Zhur. Elektr. i Tverd. Tela*, No. 27, 843-844 (1954).—It is assumed that the phase delay is caused by longitudinal electromagnetic waves propagated along the axis of the pos. column. The internal resistance of a discharge tube as a wave generator is of the order of several hundred ohms. An analysis of luminescent phenomena in the discharge shows that there is a connection between the waves and the current pulses; the periodic luminous structure disappears on lowering the pressure at a pressure P_{atm} . An equiv. circuit is developed for the discharge tube acting as a pulse generator. The amplitude and the frequency of pulsation are changing periodically with the anode-cathode distance. M. Pakaver

64

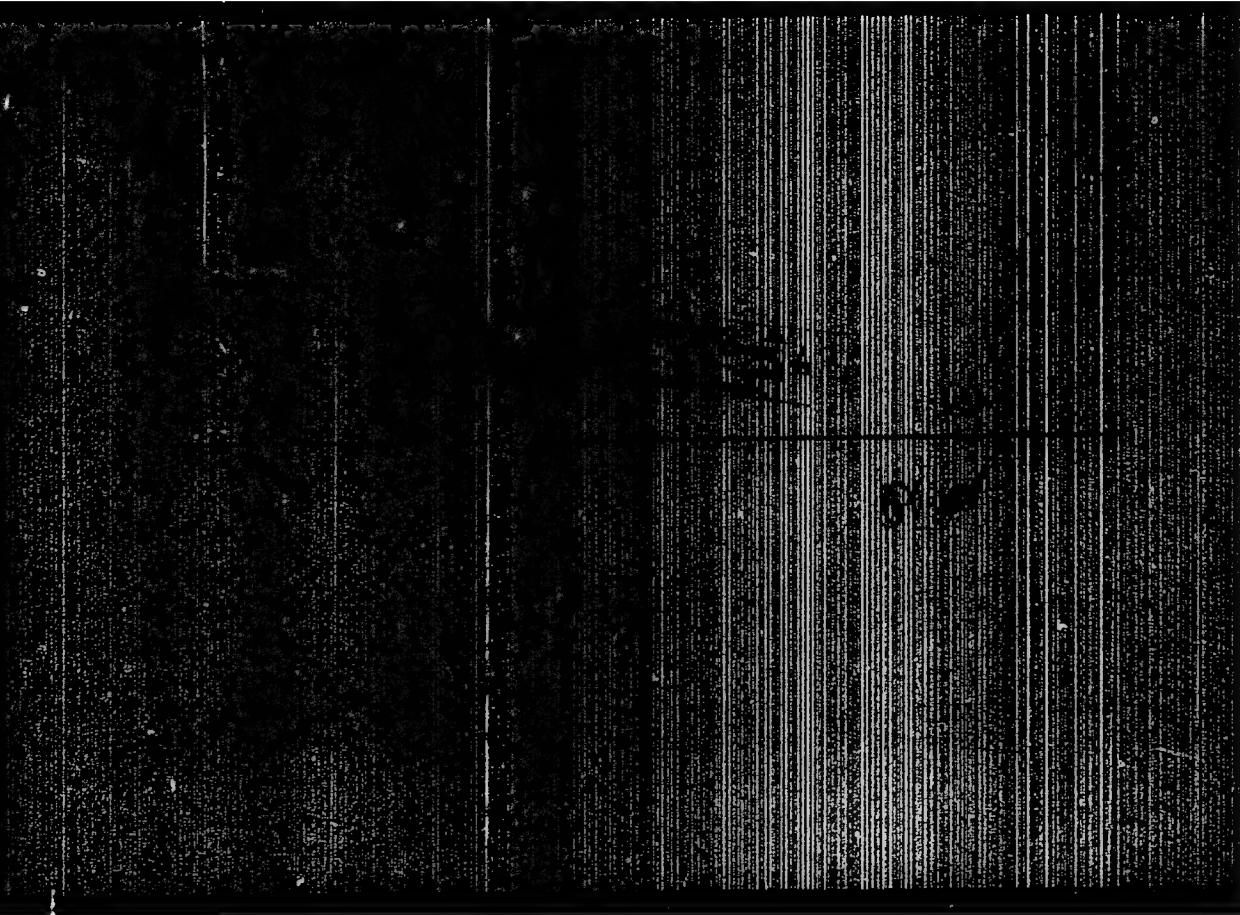
①

GERTSSENSHTEYN, M.Ye.

✓ Influence of elastic collisions of electrons and ions on longitudinal electric waves in plasma. M. B. Gerzenshteyn (Moscow State Univ.). *Zhur. Fizich. i Teor. Fiz.* 6 (2), 27, 632-8 (1964).—A simplified form is developed for the collision integral. Two cases are discussed: (1) that when the phase velocity of the longitudinal wave is large compared to thermal velocity and (2) that when the phase velocity is small. In the 2nd case the losses are so small that a small quantity of energy will cause oscillations by autocexcitation.
S. Pakishev

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

GERTSENSHTEYN, M.Ye.

Correlation of fluctuations in electron gases. Zhur. tekhn. fiz. 25
no.5:834-840 My '55. (MLRA 8:7)
(Electrons)

GERTSENSHTEYN, M.Ye.; BRYANSKIY, L.N.

Attenuator errors due to disagreement in the path of superhigh frequencies. Izm.tekh. no.1:28-33 Ja-F '56. (MLRA 9:5)
(Radio, Shortwave) (Wave guides)

GERTSENSHTEYN, M.Ye.

Determining the shunting conductivity of the probe in recording
circuits. Izm.tekh. no.4:37-38 J1-Ag '56. (MLRA 9:11)
(Electric measurements)

GERTSEVSHTEYN, M.Ye.; BRYANSKIY, L.N.

Eliminating phase distortions in power measurements. Izm. tehn.
no.6:40-43 N-D '56. (MLRA 10:1)
(Electric measurements)

AUTHOR : Gertsenshteyn, M.E. and POKRAS, A.M.

"Wave Guide Splitter with Variable Coupling."
A-U Sci Conf dedicated to "Radio Day," Moscow, 20-25 May 1957.

PERIODICAL: Radiotekhnika i Elektronika, Vol. 2, No. 4, pp. 1221-1224,
1957, (USSR)

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2

ALL INFORMATION CONTAINED
HEREIN IS UNCLASSIFIED

APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

• AUTHOR: Gertsenshteyn, V. Ye. "USSR-513-17"

TITLE: Precision Electronic Voltmeter for Relative Measurements
"Pratsizionnyy lampovyy voltmetr dlya otvetstvennykh izmereniy"

PUBL. TITLE: Izmeritel'naya tekhnika, 1955, No. 1, pp. 7-12

ABSTRACT: Measurements of **audio frequency voltages** are in most cases relative: a signal of constant frequency and form is fed to the voltmeter and the relations of the amplitudes are measured. The linearity of the amplitude characteristic is, however, insufficient. The rectifying process of the a-c **voltage** is accompanied by non-linear distortion. A new voltmeter has been developed, therefore, the circuit diagram of which is shown in Figure 1. The kenotron 6T5 is used as a rectifying tube. The amplitude value of the sinusoidal **voltage** in the grid is 50 v. The measuring circuit for checking the linearity is given in Figure 2. The results of the measurements de-

Card 1 of 2

1711-58-216-17
Precision Electronic Voltmeter for Voltage Measurements

Demonstrated that the error in the section of 10-100 divisions does not exceed 0.2%.

There are 3 graphs, 1 diagram and 1 reference, 1 of which are Bryant and German.

Card 22

IC-3-147

AUT. CR: Verte, M. N. and Bryukhav, I. N.

TITLE: *Uprugost' faze-siftera levina, a po Reflektoru (Uprugost' faze-siftera levina v odnoy volnovedayy ferevral'nosti)*PERIODICAL: Radiotekhnika i elektronika, 1978, Vol II-, No 5,
pp 710 - 721 (USSR)

ABSTRACT: The standing-wave ratio of a terminating load in a waveguide can be measured either by means of a movable probe or a fixed probe and a phase-shifter. The first method is not suitable for the measurement of small standing-wave ratios (SWR) since its accuracy is comparatively low. A higher accuracy can be achieved by employing the phase-shifter method. The equipment necessary for these measurements consists of (see Fig.1): 1) A microwave-frequency oscillator; 2) A klystron, transferred; 3) A fixed detector load; 4) A phase-shifter and, 5) the load. It is shown, however, that due to the waveguide reflections, the phase-shifter is subject to the following errors: linear errors due to the losses in the phase-shifter reflections from the movable elements of the shifter; errors due to the mis-adjusting of the oscillator and the klystron, action of the klystron. The errors due to the reflections at the elements of the phase-shifter are analysed in detail. It is proved that the phase-shifter conductors of

104-3-3-1-1/12

Waveguide Width-Shift Factor Involved in R reflection Coefficient

Dielectric waveguide thickness is a and height is h ; the permittivity of the material of the wave is ϵ and the wave number in the free space is k and expressed by:

$$\frac{a}{h} \ll 1; \quad \frac{2\pi a}{\lambda} \ll 1; \quad \lambda = \frac{\lambda_0}{\sqrt{\epsilon}} \quad (8)$$

where λ_0 is the wavelength in free space. If it is assumed that the material of the waveguide is anisotropic, the boundary conditions at the plane $x = 0$ written as Eq.(10) where \mathbf{E}^+ and \mathbf{D}^+ are the field and electric induction in the layer. The analysis of the conditions in the system can be carried out by solving Eq.(11), in which \mathbf{A} defines a vector potential. Solution of Eq.(11) is in the form of series expressed by:

$$\mathbf{A}(x, y, z) = \sum a_m(x) \mathbf{A}_m^0(x, y) \quad (14)$$

In the amplitude a_m can be obtained by solving a system of differential equations, as expressed by Eqs.(15), in which ϵ_m is given by Eq.(10). Eq.(15) can be solved by the Cardo/P method of successive approximations and in the first approximation

109-3-5-1-7-1
Waveguide Filter Reflection Coefficient

they can be expressed in the form of Eqs.(18). Solving Eqs.(19) in the form of Eqs.(20) and (21) where $\psi(z)$ is the field. On the basis of the above equations, it is possible to find the phase shift produced by the shifter if the expression by:

$$\psi = \frac{1 - \epsilon - 1}{ab} \frac{\omega}{c} \sin^2 \frac{\pi x}{a} \left\{ \begin{array}{l} \text{in air} \\ \text{at } z = \infty \end{array} \right. \quad (27)$$

where a and b are the dimensions of the waveguide and x is the distance between the plate of the phase-shifter and the narrow wall of the guide. The reflection coefficient of the phase-shifter can be expressed by:

$$- \frac{1}{ab} \frac{\omega}{c} \sin^2 \frac{\pi x}{a} \left\{ \frac{\epsilon - 1}{\sqrt{1 - \omega_0^2/c^2}} \right\} \text{Re}(\psi) e^{-2iz} dz \quad (28)$$

which, for a symmetrical plate, is in the form of Eq.(19). Eqs.(27) and (28) can be regarded as the basic formulae for Card 5/7

100-3-5-147

Waveguide Phase-shifter Having a Low Reflection Coefficient

the design of a phase-shifter. It is shown that the error of measurement of the reflection coefficient of the load $\tilde{\Gamma}_L$ is equal to the reflection coefficient of the phase-shifter, $\tilde{\Gamma}_Q$, by means of Eq.(30). From this, it follows that the error conditions (maxim error) are expressed by:

$$\tilde{\Gamma}_h = \frac{\sqrt{2}}{2} \tilde{\Gamma}_Q = 0.407 \tilde{\Gamma}_Q \quad (31).$$

The reflection coefficient of a phase-shifter can be measured experimentally by means of the equipment shown in Fig.7; this consists of a wave detector load, an auxiliary phase-shifter, a compensated phase-shifter, a matching transformer and a terminated load. Eq.(29) can be used to design a phase-shifter $\tilde{\Gamma}_Q$, i.e. Eq.(29) is transformed into Eq.(31), provided that $\tilde{\Gamma}_L$ is given by Eq.(30). In this equation, (μ_2) denotes the average per dimensions of the phase-shifter $\tilde{\Gamma}_Q$ at its largest circumscription (in the centre). Eq.(38) shows that $\tilde{\Gamma}_Q = 0.407 \tilde{\Gamma}_L$ is a satisfactory choice of the phase-shifter $\tilde{\Gamma}_Q$, since $\tilde{\Gamma}_Q$ is given by Eq.(40), where α is a parameter. Optimum $\tilde{\Gamma}_Q$ is given by Eq.(41) in the case of a shorted load.

100-3-5-1-1/1
W vegetable Phase-shifter Pave in Low Reflective Coatings

The form of Eq.(41). An experimental phase-shifter, based on Eq.(41), was constructed and it was found that its reflectance of incident was so low that it could not be measured directly, without a dummy line. It was found by employing the method of Fig. 5 that the standing wave ratio was better than 1.00. There are 7 figures and 12 references, 9 of which are Soviet and 3 English.

ASSOCIATION: Vsesoyuznyy n.-i institut fiziko-tehnicheskikh i radio-tehnicheskikh izmereniy (All-Union Scientific Research Institute for Physico-engineering and Radio-engineering Measurements)

SUBMITTED: July 30, 1956

AVAILABLE: Library of Congress
Card F/S

1. Wave ratio-Measurement 2. Phase shifter-Applications

AUTHOR: Gertsenshteyn, M. Ye.

10/10/86-10-7/1

TITLE: Spatial Beats of noise Waves in Coupled Delay Devices
(Lines) (Prostranstvennyye liyeniya shukovikh voln v
svyazannykh zamedlitelyakh)PERIODICAL: Radiotekhnika i Sistemika, 1986, vol 3, no 10,
pp 1254 - 1263 (USSR)ABSTRACT: The investigation of complex problems of wave propagation in electron beams or in electron gas can be conveniently treated as a problem of formal electrodynamics, since the fundamental equations contain a permittivity operator $\hat{\epsilon}$ for the electron gas. The Maxwell equations are therefore written in the form of Eqs.(1) where j and ρ are the currents and charges which excite the waves. The permittivity is a function of frequency ω and of the wave number k , i.e:

$$\hat{\epsilon} = \hat{\epsilon}(\omega, k) \quad (1)$$

Here the sign on top denotes an operator. It is seen the electrodynamic equations, similar to classical, arise; firstly, the system (electron gas and the electromagnetic field) have a large number of degrees of

Sov/100-3-10-3/12
Spatial Beats of Noise Waves in Coupled Delay Devices (11.6m)

freedom; secondly, the meaning of \mathbf{j} and ρ is not clear and the sources of noise are not taken into account. It is therefore necessary to consider the following kinetic equation for the distribution function f :

$$\frac{\partial f}{\partial t} + \mathbf{v} \cdot \frac{\partial f}{\partial \mathbf{r}} + \frac{e}{m} \left\{ \mathbf{E} + \left[\frac{\mathbf{v} \times \mathbf{H}}{c} \right] \right\} \frac{\partial f}{\partial \mathbf{v}} = 0 \quad (4).$$

The electromagnetic fields have to satisfy the equation system (4). The oscillatory component of the distribution function φ is expressed by Eq.(5), where $\hat{\mathbf{L}}$ is a linear operator, as defined by Eq.(3). From the theory of linear, differential equations (Ref 9), it follows that the solution of Eq.(5) in a fixed region of space can be represented as Eq.(7), where φ_{sv} corresponds to free oscillations and is independent of the field, while φ_{syn} is analogous to the forced oscillations and proportional to the right-hand side of Eq.(5).

Card2/6

General Solution of the Poisson's Equation. (11)

Consequently, the second component of Eq.(11), the boundary conditions given in Eq.(5), describes the motion of electrons under the influence of the three components of the electromagnetic field. The solution of Eqs.(1) can be written as Eq.(12), where \hat{G} is the Green function of Eq.(5). The third and components of the charges and currents can be expressed in Eq.(11), as the component j_{vn} can be linearly expressed by the field E , as shown in Eq.(12). The operator \hat{G} is given by Eq.(13). The shot noise in electron noise can be evaluated on the basis of the distribution function given by Eq.(15), in which r_i and v_i are the radius vector and velocity of the i -th electron. On the other hand, the component of the distribution function, described by Eq.(3), which describes the noise, is expressed in Eq.(16). Consequently, the function ϕ_{sh} can be expressed by analogy with Eq.(7), as a sum of two components (Eq.(17)). The current and the charge noise can be obtained in Eqs.(18), where β_i is the noise current in Eq.(16).

0-163/6

SCV/27-7-10-2/1
2. Field Equations of Noise Model. Coupled Delay Devices (11)

electron. On the basis of the above analysis, it is concluded that an arbitrary linear system can be described by the following equations of the noise in the coupled dynamics:

$$\begin{aligned} \text{rot } \mathbf{H} &= \frac{i_0}{c} \hat{\epsilon} \mathbf{E} + \frac{4\pi}{c} (j_A + j_{sv} + j_{sh}) ; \\ \text{rot } \hat{\mathbf{E}} &= - \frac{i_0}{c} \mathbf{H} ; \\ \text{div } \hat{\epsilon} \mathbf{E} &= 4\pi (\rho_A + \rho_{sv} + \rho_{sh}) ; \\ \text{div } \mathbf{H} &= 0 , \end{aligned} \quad (11)$$

where j_A and ρ_A are the currents and charges in the antennae which excite the system. The solution of the normal wave (the solution of Eqs.(11)) with the form

Con 54/6

3.3.3. The slow-wave \mathbf{H} in Cavity \mathbf{H}_1 of Divided (II)

In slow-wave system, the \mathbf{j}_1 is the current density, \mathbf{j}_2 is the wave \mathbf{j}_2 is express by Eq.(21), current \mathbf{j}_1 and \mathbf{j}_2 are very weak, the wave transmission coefficient is very small (waveguide). In the slow-wave device (slow-wave system) contains only one wave, the amplitude of the wave is given by Eq.(24), where β_0 is the wave vector of the device, e is the electron charge and γ is the normalized coefficient dependent on the structure of the field in the device. The solution of the equation has the form of Eq.(30), where k_3 is the amplification coefficient, which is expressed by the last expression in Eq.(30). From this, it is seen that an exchange of energy between the two waves. By analogy to the optical slow-wave system, these phenomena can be referred to as the optical noise beats in slow-wave systems. The noise beats in optical systems were first discovered by Krasnopol'skii and Lev (Ref.1) and today they find application in optics.

On: d5/6

1. V/11-5-24-4
Spatial Beats of Noise Waves in Coupled Delay Devices (Eng.)

There are 17 references, 15 of which are Soviet, 1 English and 1 German; three of the Soviet references are translated from English.

ASSOCIATION: Vsesoyuznyy n.-i in-t fiziko-tekhnicheskikh i radiotekhnicheskikh issledovanii (All-Union Scientific Research Institute of Physical-technical and Radio-engineering Measurements)

SUBMITTED: July 30, 1956

Card 6/6 1. Delay lines--Theory 2. Electromagnetic fields--Mathematical analysis

6(4), 7(7)

SC7-105-13-12-3/12

AUTHORS:

Gertsenshteyn, M. Ye., Pokras, A. N., Dol'skay, L. G.

TITLE:

Multi-Channel System of Parallel Selection Waveguides with
Variable Couplings (Mnogostvol'nyaya sistema parallel'noy
seleksii s reguliruyemymi svyaz'ami)

PERIODICAL:

Radiotekhnika, 1958, Vol 13, Nr 12, pp 20-25 (USA)

ABSTRACT:

With relatively narrow bands or not too high claims with respect to the adaptation, the problem of dividing or joining the channels can be solved by means of a system of shunted series-resonance circuits. The various filters are connected, in parallel to each other, to the common conductor by a simple or compact tap. A simple method of setting up a tap for the shunted series-resonance circuits is given. This method is based on the calculation data without intricate experimental work. At first, the paralleling of the resonance circuits is investigated. The obtained formulae (3) and (5) show that the tap must be tuned jointly with the filter connected to it, with one element. The input resistance of filters with several elements is then investigated and it is shown that the mutual influence of the various channels is determined essentially by the input resona-

Card 1/2

77-108-13-12-37-12

Multi-Channel System of Parallel Selection Waveguides with
Variable Couplings

tors. Therefore, the input resonators of the filters with several elements must also be tuned with the taps. The connection of the filters to the common line is then investigated. The connection to the main waveguide is made variable by means of screws with a steplike cross section. By means of the method given in this article, a simple waveguide tap is worked out for a system with shunted series-resonance circuits with an input transient wave factor of ≈ 0.95 in the middle of the band. There are 7 figures, 1 table, and 3 Soviet references.

SUBMITTED: June 1, 1957

Card 2/2

GERTSENSHTEYN, M. YE.

56-1-55/56

AUTHORS: Bonch-Bruyevich, V. L. , Gertsenshteyn, M. Ye.

TITLE: On the Theory of the Magnetic Susceptibility of Metals (K teorii magnitnoy vospriimchivosti metallov)

PERIODICAL: Zhurnal Eksperimental'noy i Teoreticheskoy Fiziki, 1958, Vol. 34, Nr 1, pp. 261 - 261 (USSR)

ABSTRACT: The magnetic susceptibility of the electron gas was recently (references 1, 2, 3) calculated with the taking into account of the distant Coulomb correlation. In this connection, however, only the susceptibility caused by the Fermi branch of the spectrum of excitations was taken into account. But the authors want to call attention to the fact that the Bose quanta of plasma vibrations also furnish a certain contribution to the susceptibility. It is true that these excitations are neutral and do not furnish any contribution to the current, but their energy depends on the field strength of the magnetic field H and therefore the plasma-quanta are "carriers of magnetism". At the usual temperatures the real plasma-quanta are practically not excited in metal, but their zero energy also depends on H . This leads, as shown here, to a plasma-diamagnetism comparable with the Landau diamagnetism. In a weak magnetic field a separation of the plasma vibrations in longitu-

Card 1/2

56-1-55/56

On the Theory of the Magnetic Susceptibility of Metals

dinal and transversal vibrations is also possible. For the case discussed here only the former are of interest. An expression for the frequency of the longitudinal plasma-quantum is given. Then the author gives an expression for the magnetic susceptibility caused by the dependence of the zero energy of the plasma on the magnetic field. The neglect of the zero energy of the plasma is generally not at all justified and the quantitative agreement of the theory by Pines (reference 1) with the experiment must anew be checked. There are 5 references, 2 of which are Slavic.

ASSOCIATION: **Moscow State University**
(Moskovskiy Gosudarstvennyy universitet)

SUBMITTED: November 21, 1957

AVAILABLE: Library of Congress

Card 2/2

~~CONFIDENTIAL~~ 21. Yes

2. N. Kurnov

Широко распространенные методы определения времени
дней

3. A. Gorobec

О методах измерения в радиотехнической гидро-
акустике в частотной подсистеме

18 часов

(с 18 до 22 часов)

4. N. Prost

Измерение времени работы в температуре до
теплопроводности, а также о методах определения
изменения времени

5. N. Kurnov

О методах определения времени

6. E. Gorobec

7. E. Kudryavtsev

Факторы, влияющие на температуру времени
изменения

8. N. Danov

О методах измерения в радиотехнической гидро-
акустике в частотной подсистеме

9.

10. N. Kurnov

О гидроакустическом времени в измерении
изменения времени

11. hours

(с 10 до 16 часов)

12. N. Kurnov

Изменение времени измерения в гидро-
акустике

13. N. Kurnov

Изменение времени измерения

14. N. Danov

О методах определения времени измерения

15. hours

(с 10 до 22 часов)

67

report submitted for the Centennial Meeting of the Scientific Technological Society of
Radio Engineering and Electrical Communications Inc. A. S. Popov (TEKHN), Moscow,
6-10 June, 1959

AUTHOR: Gertsenshteyn, M.Ye.

007/100-4-1-a7/50

TITLE Noise in an Electron Beam (O shumakh elektronnoy puchki)PERIODICAL. Radiotekhnika i Elektronika, 1964, Vol 4, Nr 1,
pp 146 - 147 (USSR)

ABSTRACT An electron beam contains two types of noise; one of these can be referred to as the cathode noise and is due to the emission processes at the cathode which produces the beam. The second type of noise can be referred to as the volume noise and is due to the processes occurring in the electron beam itself. In the vicinity of the cathode, the cathode noise is predominant while the volume noise is comparatively low. It can be expected that at large distances from the cathode, the volume noise will become significant, while the cathode noise is negligible. It is shown that the conditions for the predominance of the volume noise can be expressed by.

$$\frac{e}{m} \cdot \frac{1}{r} \cdot 0.3 - 0.4 \quad (6)$$

$$\omega_0 \cdot \frac{1}{r} \geq 0.2 - 0.5 \quad (7)$$

CIA-RDP86-00513R000514920017-2

Noise in an Electron Beam

307/109-4-1-21/30

where ζ is given by Eq (5), τ is the transit time for the drift space and ω_c is the Langmuir frequency.

In Eq (5), u_0 is the electron beam velocity, v_0 is the thermal electron velocity and ω is the operating frequency.

There are 3 references, 2 of which are Soviet and 1 English.

SUBMITTED. February 21, 1958

Card 2/2

16(1),16(2)

AUTHORS: Gertsenshteyn, M Ye., and Vasil'yev, V B. 05/93
30V/52-4-4 4/13

TITLE: Waveguide With the Random Inhomogeneities and Brownian Motion
on the Lobachevskiy Plane

PERIODICAL: Teoriya veroyatnostey i yeye primeneniya 1958,
Vol 4, Nr 4, pp 424-432 (USSR)

ABSTRACT: The authors consider a waveguide with random inhomogeneities. Let r_1 be the reflection coefficient (ratio of the amplitudes of the reflected and original wave) of a single inhomogeneity. Let all r_i be independent random functions with known statistical characteristics. The authors ask for the reflection coefficient of the whole waveguide. It is shown that the problem can be reduced to the Brownian motion in the Lobachevskiy plane. At first two inhomogeneities are considered and it is stated that the resulting reflection coefficient is a bounded linear function mapping the unit circle onto itself. Thencewith the relation with the Lobachevskiy plane is given. For several inhomogeneities the image point moves in the Lobachevskiy plane, while the sum of the corresponding normalized distances yields the total effect of the inhomogeneities. If the considered random process is continuous, then it leads to the diffusion equation in the Lobachevskiy plane.

SUBMITTED: December 25, 1958
Card 1/1

AUTHORS: Gertsenshteyn, M.Ye. and Vasil'yev, V.B. SOV/109-4-4-7/24

TITLE: The Diffusion Equation of a Statistically Non-homogeneous Waveguide (Diffuzionnoye uravneniye dlya statisticheski neodnorodnogo volnovoda)

PERIODICAL: Radiotekhnika i elektronika, 1959, Vol 4, Nr 4, pp 611 ~ 617 (USSR)

ABSTRACT: It is assumed that the complex reflection coefficient of the system is $r = x + iy$ and that its probability density distribution satisfies the diffusion equation:

$$D \left(\frac{\partial^2 W}{\partial x^2} + \frac{\partial^2 W}{\partial y^2} \right) = - \frac{\partial W}{\partial z} \quad (3)$$

where D is the statistical characteristic of the waveguide; this is equal to the average half sum of the reflection coefficients squared per unit length of the waveguide; z is the distance along the length of the waveguide. If a normalised variable $t = \int D dz$ is introduced, the equation can be written as Eq (4). When the waveguide

Card1/4

SOV/10 - 1-4-7/24
The Diffusion Equation of a Statistically Non-homogeneous Waveguide

is terminated with a matched load, the solution of Eq (4) is in the form of Eq (5). It is seen that for large t , Eq (5) has no physical meaning. A different differential equation for the density probability function is, therefore, necessary. The equation should be such as to make the solution independent of the terminating load; also when $x^2 + y^2 \neq 0$, the differential equation should coincide with Eq (4). These requirements are satisfied by

$$\Delta W - \frac{\partial W}{\partial t} \quad (7)$$

where Δ is the Laplace operator on the Lobachevskiy surface. The operator is defined by Eq (8). By introducing a polar system of co-ordinates η, ϕ , as defined by Eqs (9), the Laplace operator is represented by Eq (10). If $\eta = i\theta$ and $u = \text{ch} \eta$, Eq (10) can be expressed as Eq (11). This can be solved by introducing the Laplace transformations and leads to the Legendre equation which

Card2/4

SOV/109-4-4-7/24

The Diffusion Equation of a Statistically Non-homogeneous Waveguide

is in the form of Eq (15). In its final form, Eq (15) can be written as Eq (16). On the basis of the above, it is found that the average value for u is expressed by Eq (17). The average value of the reflection coefficient is approximately expressed by Eq (19). The value of the average reflection coefficient r as a function of t is plotted in Figure 2; Curve I corresponds to a linear approximation, while Curve II represents more accurate results. It is seen that Curve I gives values which are higher than those represented by Curve II. The physical meaning of this is that a part of the energy of the reflected wave travelling from the load towards the generator is reflected by the non-uniformities of the waveguide (towards the terminating load). The authors make acknowledgment to B.Ye. Kinber for discussing the work and for his valuable remarks.

Card 5/4

SOV/109-4-4-7/24

The Diffusion Equation of a Statistically Non-homogeneous Waveguide

There are 2 figures and 9 references, 1 of which is English and 8 Soviet. 1 of the Soviet references is translated from English.

SUBMITTED: November 26, 1957

Card 4/4

REF ID: A6570

367

AUTHORS: Polubarnov, I. I. and Gorbunov, Yu. N.

TITLE: Possibility of Measuring the Value of Gravity by a Gyrostat under Laboratory Conditions

PERIODICAL: Zhurnal experimentalnoi i teoreticheskoy fiziki (JETP)
Year: 57, No. 6, pt. 2 (1969)

ABSTRACT: A theoretical analysis was made of the possibility of experiment which would measure the value of gravity on the proposition of a gyroscope with a rotating gravitational field and of the possibility of measuring the value of gravity by a rotating wheel with a rapidly rotating axis of rotation with nearly constant angular velocity. The theory of the gyroscope with a rotating gravitational field was developed by Newton using potential. The stability of the gyroscope with a rotating gravitational field was analyzed in the plane of the wheel. The theory of the gyroscope with a rotating gravitational field was developed by using the property of rotation of the gyroscope about the axis of the wheel during the rotation of the wheel around the axis of rotation.

Card 1/3

Possibility of $Y = 0$ is $1 - 7 \times 10^{-7}$.
Probability of $Y = 1$ is 7×10^{-7} .
Probability of $Y = 2$ is 10^{-7} .

2. *Observe the following effect:* $\text{Na}_2\text{O}_2 + 2\text{H}_2\text{O} \rightarrow 2\text{NaOH} + \text{O}_2$

1100

在於此，故其後人之學，亦復以爲子思之學，而不知子思之學，實爲孟子之學也。

Chap. 1

Possibility of Measuring the Velocity of
Gravitational Distribution under Laboratory
Conditions

770.4
SOW/5C-37-6-53, 11

gravity. There is a Scott reference.

SUBMITTED: July 29, 1954

Card 3/3

GERTSENSHTEYN, M. Ye.; BRYANSKIY, L.N.

Using phase shifters for eliminating mismatch errors. Izm.tekh.
no.1:48-51 Ja '60. (MIRA 13:5)
(Phase converters)

$\frac{1}{\mu} \mathbf{E} \times \mathbf{B} = \mathbf{J}$

μ

$\mathbf{E} \times \mathbf{B} = \mu \mathbf{J}$

$\text{rot } \mathbf{H} = \frac{1}{\mu} \mathbf{D} - \frac{1}{\mu} \mathbf{J}$

$\text{rot } \mathbf{H} = \frac{1}{\mu} \mathbf{D} - \frac{1}{\mu} \mathbf{J}$

$\text{rot } \mathbf{E} = \frac{1}{\mu} \mathbf{B} - \mathbf{0}$

Consequently, the equations of motion reduce to
Gauge invariant form, namely, Eq. (17) and
Variable ρ is given by

$$\mathbf{D} = \varepsilon(t) \mathbf{E}_t \quad (1)$$

$$\mathbf{B} = \mu(t) \mathbf{H}_t \quad (2)$$

\mathbf{E} and \mathbf{H} are weak at $t = 0$ and the system is
approximately uncoupled. The system is
approximately uncoupled. The system is
approximately uncoupled.

$$= \frac{e}{4\pi} \operatorname{div}[\mathbf{EH}] + \frac{1}{8\pi} \frac{d}{dt} (\varepsilon \mathbf{E}, \mathbf{E}) + \mu \mathbf{H}_t \cdot \mathbf{H}_t = \frac{1}{2} \varepsilon \mathbf{E}^2 - \mathbf{H}_t \cdot \mathbf{H}_t = 0$$

which implies that \mathbf{E} and \mathbf{H} are approximately
uncoupled. The system is approximately
uncoupled. The system is approximately
uncoupled.

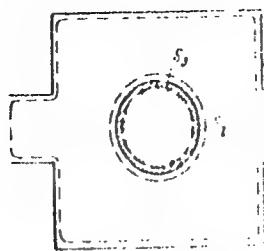


Diagram illustrating the relationship between the area of the central circular component (S1) and the area of the entire rectangular frame (S2). The central component has a wavy outer boundary. The frame has a cross-shaped cutout in the center. The area of the central component is labeled S1, and the area of the entire frame is labeled S2.

$$P_1 = \frac{1}{2} \int_{S1} (E\dot{H}) d\sigma$$

$$P_2 = \frac{1}{2} \int_{S2} (E\dot{H}) d\sigma$$

Consequently, Electromagnetic field energy in the cavity is given by the following: A Hrotograph (H) in which \mathbf{E} and \mathbf{H} are the electric and magnetic field parameters

$$W = \frac{1}{2} \epsilon_0 \int \mathbf{E} \cdot \mathbf{D} + \frac{1}{2\mu_0} \int \mathbf{H} \cdot \mathbf{B}$$

$$P_e - P_h = \frac{d}{dt} \frac{1}{8\pi} \int (\epsilon \mathbf{E}_i \cdot \mathbf{E}_i + \mu \mathbf{H}_i \cdot \mathbf{H}_i) dr + \frac{1}{8\pi} \int (\mathbf{E}_i \cdot \mathbf{E}_i + \mu \mathbf{H}_i \cdot \mathbf{H}_i) dr - \epsilon \mathbf{E}_i \cdot \mathbf{H}_i$$

The above is a rate of change of total energy in the cavity with respect to time. While (\cdot) is obtained by integration of \mathbf{E} and \mathbf{H} which are unknown to us, $(\cdot)_i$ is the energy in the cavity obtained by an approximate method based on the structure. Relationship between (\cdot) and $(\cdot)_i$ can be obtained: for one-dimensional case, the energy in the cavity with self-consistent \mathbf{E} and \mathbf{H} is given by the following: ϵ and μ are the dielectric constants of normal incident plane waves in the one-dimensional medium. The incident wave is a function of a resonator profile, $\epsilon(x)$ and $\mu(x)$ are the constant wave $\epsilon = \epsilon_0(\epsilon')$ and $\mu = \mu_0(\mu')$. The incident wave is given by $\mathbf{E} = E_0 \sin(kx)$ and $\mathbf{H} = H_0 \sin(kx)$. The energy in the cavity is given by:

Let \mathbf{A} be a 3×3 matrix with entries A_{ij} and let \mathbf{M} be a 3×3 matrix with entries M_{ij} .

$$\mathbf{e} = \mathbf{e}^0 + \mathbf{e}_1(t),$$

$$\mathbf{p} = \mathbf{p}^0 + \mathbf{p}_1(t).$$

Let \mathbf{E} and \mathbf{H} be the electric and magnetic fields respectively. Then \mathbf{E} and \mathbf{H} are given by the following equations:

$$\mathbf{E} = \sum_{\mathbf{k}} a_{\mathbf{k}}(t) \mathbf{E}_{\mathbf{k}}(\mathbf{r}),$$

$$\mathbf{H} = \sum_{\mathbf{k}} b_{\mathbf{k}}(t) \mathbf{H}_{\mathbf{k}}(\mathbf{r}).$$

Let \mathbf{E}_q be the electric field due to the charge q and let \mathbf{H}_q be the magnetic field due to the charge q . Then \mathbf{E}_q and \mathbf{H}_q are given by the following equations:

$$\text{rot } \mathbf{E}_q = -\frac{1}{c} \omega_q \mathbf{z}^* \mathbf{H}_q,$$

$$\text{rot } \mathbf{H}_q = \frac{1}{c} \omega_q \mathbf{z}^* \mathbf{E}_q.$$

3 and 6/11

$$\begin{aligned}
& \left(\frac{1}{1 - \lambda_1 z} + \frac{1}{1 - \lambda_2 z} + \cdots + \frac{1}{1 - \lambda_n z} \right) \left(\frac{1}{1 - \mu_1 z} + \frac{1}{1 - \mu_2 z} + \cdots + \frac{1}{1 - \mu_n z} \right) \\
& = \left(\sum_{q=1}^n a_q b_q z^q \mathbf{E}_q \right) \left(\sum_{q=1}^n a_q b_q z^q \mathbf{E}_q \right) = \sum_{q=1}^n \left[(a_q z^q + \frac{1}{\lambda_1} z^{q-1} + \cdots + \frac{1}{\lambda_n} z^{q-n}) (b_q z^q + \frac{1}{\mu_1} z^{q-1} + \cdots + \frac{1}{\mu_n} z^{q-n}) \right] \\
& = \sum_{q=1}^n a_q b_q z^q \mathbf{H}_q = \sum_{q=1}^n \left[(a_q z^q + \frac{1}{\lambda_1} z^{q-1} + \cdots + \frac{1}{\lambda_n} z^{q-n}) (b_q z^q + \frac{1}{\mu_1} z^{q-1} + \cdots + \frac{1}{\mu_n} z^{q-n}) \right] \quad (14)
\end{aligned}$$

a)

$$\begin{aligned} V_1(t) &= \int_{\mathbb{R}^3} \mathbf{E}_1(\mathbf{r}, t) \cdot \mathbf{E}_1(\mathbf{r}, t) d\mathbf{r} \\ &= \int_{\mathbb{R}^3} \mathbf{H}_1 \mathbf{g}_1(\mathbf{r}, t) \mathbf{H}_1 \mathbf{g}_1(\mathbf{r}, t)^T d\mathbf{r} = \mathbf{g}_1^T \mathbf{H}_1 \mathbf{g}_1 \end{aligned}$$

and the second term in the right hand side of (1) is

the energy of the system. The energy of the system is the sum of the kinetic energy of the system and the potential energy of the system. The potential energy of the system is the sum of the potential energy of the particles and the potential energy of the fields.

$$\begin{aligned} \partial_t \mathbf{g}_1 &= \frac{d}{dt} \left[(1 + \beta) \frac{1}{2} \mathbf{g}_1^T \mathbf{H}_1 \mathbf{g}_1 + \mathbf{g}_1^T \mathbf{H}_1 \mathbf{g}_1 \right] \\ &= \partial_t \mathbf{g}_1 + \frac{1}{2} \left[(1 + \beta) \frac{1}{2} \mathbf{g}_1^T \mathbf{H}_1 \mathbf{g}_1 + \mathbf{g}_1^T \mathbf{H}_1 \mathbf{g}_1 \right] \end{aligned} \quad (17)$$

Concerning Electromagnetic Waves
Containing A Gradient Media With
Variable Parameters

In a similar way, keep that the wave function can be transformed to the wave function ϵ_{α}^t and μ_{α}^t only. (1) we get the wave function of the normal waves in gradient media ϵ_{α}^t and μ_{α}^t (ω, t), $\mu_{\alpha}^t = \mu(\omega, t)$ in the following form:

$$\epsilon_{\alpha}^t(\omega, t) = \sum_{m=0}^{+\infty} \epsilon_m^t(m) e^{i\Omega_m t},$$

$$\mu_{\alpha}^t(\omega, t) = \sum_{m=0}^{+\infty} \mu_m^t(m) e^{i\Omega_m t},$$

$$\begin{aligned} \text{Re}(\omega) &= \omega_0 - \frac{\epsilon_0 - \mu_0}{2\Omega_0}, & \text{Im}(\omega) &= \frac{\epsilon_0 - \mu_0}{2\Omega_0}, \\ \text{Re}(\Omega) &= \Omega_0, & \text{Im}(\Omega) &= \frac{\epsilon_0 - \mu_0}{2\omega_0}, \\ \text{Im}(\omega) &= \frac{\epsilon_0 - \mu_0}{2\Omega_0}, & \text{Im}(\Omega) &= \frac{\epsilon_0 - \mu_0}{2\omega_0}, \end{aligned}$$

where ω_0 is the resonance frequency of the wave.

ω

E-D B H

E H

$$E = \sum_n E_n e^{i\omega_n t}$$

$$H = \sum_n H_n e^{i\omega_n t}$$

and $\langle \hat{H} \rangle = \langle H \rangle$ since $\langle \hat{H}^2 \rangle = \langle H^2 \rangle$

$$\overline{\text{div}}[EH^*] = \frac{1}{i} \sum_n \sum_m m_n (E_n - E_m) \langle H_n | H_m \rangle = \langle H^* \rangle - \langle H \rangle$$

and $\langle \hat{H}^2 \rangle = \langle H^2 \rangle$ since $\langle \hat{H}^3 \rangle = \langle H^3 \rangle$

Hence $\langle \hat{H}^2 \rangle$

the second term in the expression for \mathcal{L}_0 is zero, since \mathcal{L}_0 is a linear operator. A similar argument applies to the third term.

$$\begin{aligned} i \sum_j b_j \phi_j Q_{ij}^{(m)}(u) &= \sum_j b_j Q_{ij}^{(m)}(u) = \sum_{j=1}^d b_j \delta_{ij} (u) = b_i, \\ i \sum_j a_j \phi_j Q_{ij}^{(m)}(u) &= \sum_j a_j Q_{ij}^{(m)}(u) + \sum_j a_j (c_j, f_j) \in U. \end{aligned} \quad (17)$$

Thus, $\varphi_{\mu_0, \omega_0, 1} = \psi_{\mu_0, \omega_0, 1}$ and the proof is complete.

From the above, we have $\varphi_{\mu_0, \omega_0, 1} = \psi_{\mu_0, \omega_0, 1}$ and the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, $\omega_0 = \delta_{ij} \phi_j Q_{ij}^{(m)}(u)$ and the proof is complete.

Since the expression for \mathcal{L}_0 is linear, the proof is complete.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

Therefore, the operator ω_0 of the second term in the expression for \mathcal{L}_0 is a linear operator.

$$\frac{\partial^2 \Psi}{\partial x^2} + \frac{\partial^2 \Psi}{\partial y^2} + \frac{\partial^2 \Psi}{\partial z^2} = 0$$

Let $\Psi = \Psi(x, y, z)$ be a function of x, y, z such that Ψ is a solution of the Laplace's equation. Then Ψ is called a harmonic function.

$$\Psi: \mathbb{H} \rightarrow \text{grad } \Psi$$

$$\frac{\partial \Psi}{\partial x} = \frac{\partial \Psi}{\partial y} = \frac{\partial \Psi}{\partial z} = 0$$

$$\text{... div grad } \Psi = \frac{\partial^2 \Psi}{\partial x^2} + \frac{\partial^2 \Psi}{\partial y^2} + \frac{\partial^2 \Psi}{\partial z^2} = 0$$

$$\frac{\partial^2 \Psi}{\partial x^2} + \frac{\partial^2 \Psi}{\partial y^2} + \frac{\partial^2 \Psi}{\partial z^2} = 0$$

$$\frac{\partial^2 \Psi}{\partial x^2} + (1+k) \left(\frac{\partial^2 \Psi}{\partial y^2} + \frac{\partial^2 \Psi}{\partial z^2} \right) = 0$$

where μ_1 is the chemical potential of the system.

For simplicity,

$$\Delta W = \frac{\partial W}{\partial \mu} + \frac{\partial W}{\partial \epsilon} - \frac{\partial W}{\partial \mu} = 0$$

is assumed. Then, the chemical potential is given by

from $\mu_1 = \mu_1(\epsilon, T, \mu_0)$ (where μ_0 is the chemical potential of the system at $\epsilon = 0$ and $T = 0$),

and the energy of the system is given by

from $W = \int d\mathbf{r} \epsilon \delta(\mathbf{r}) \delta(\mathbf{r})^2 \exp(-\epsilon/\mu_1)$ (where $\delta(\mathbf{r})$ is the δ -function).

Therefore, the energy of the system is given by

from $\epsilon > \mu_1$ and $\epsilon < \mu_1$,

from $\epsilon = \mu_1$ and $\epsilon \neq \mu_1$,

$$S = \frac{c}{4\pi} [\mathbf{E} \mathbf{H}^*] + \frac{m^2 J_1}{8\pi^2 k} \mathbf{H} \mathbf{H}^*$$

where \mathbf{k} is the wave vector, \mathbf{E} is the electric field, \mathbf{H} is the magnetic field, c is the speed of light, m is the mass of the particle, J_1 is the current density, and k is the Boltzmann constant.

From the above, the energy of the system is given by

from $\epsilon > \mu_1$ and $\epsilon < \mu_1$,

from $\epsilon = \mu_1$ and $\epsilon \neq \mu_1$,

1. (b) (5) (A) (ii) (B) (1) (C) (1) (D) (1) (E) (1) (F) (1) (G) (1) (H) (1) (I) (1) (J) (1) (K) (1) (L) (1) (M) (1) (N) (1) (O) (1) (P) (1) (Q) (1) (R) (1) (S) (1) (T) (1) (U) (1) (V) (1) (W) (1) (X) (1) (Y) (1) (Z)

2. (b) (5) (A) (ii) (B) (1) (C) (1) (D) (1) (E) (1) (F) (1) (G) (1) (H) (1) (I) (1) (J) (1) (K) (1) (L) (1) (M) (1) (N) (1) (O) (1) (P) (1) (Q) (1) (R) (1) (S) (1) (T) (1) (U) (1) (V) (1) (W) (1) (X) (1) (Y) (1) (Z)

3. (b) (5) (A) (ii) (B) (1) (C) (1) (D) (1) (E) (1) (F) (1) (G) (1) (H) (1) (I) (1) (J) (1) (K) (1) (L) (1) (M) (1) (N) (1) (O) (1) (P) (1) (Q) (1) (R) (1) (S) (1) (T) (1) (U) (1) (V) (1) (W) (1) (X) (1) (Y) (1) (Z)

4. (b) (5) (A) (ii) (B) (1) (C) (1) (D) (1) (E) (1) (F) (1) (G) (1) (H) (1) (I) (1) (J) (1) (K) (1) (L) (1) (M) (1) (N) (1) (O) (1) (P) (1) (Q) (1) (R) (1) (S) (1) (T) (1) (U) (1) (V) (1) (W) (1) (X) (1) (Y) (1) (Z)

9.4600,9.2180

77772
SOV/109-5-2-5/26

AUTHOR: Gertsenshteyn, M. E.

TITLE: Phase and Frequency Distortions in Mixers

PERIODICAL: Radiotekhnika i elektronika, 1960, Vol 5, Nr 2,
pp 214-217 (USSR)

ABSTRACT: Amplitude and phase distortions in crystal mixers
at super high frequencies are analyzed assuming
that the mixer is a six-pole network which can be
described by a corresponding matrix of conductivity.
This leads, however, to cumbersome calculations and
not comprehensive end results. Provided the non-
uniformity of the frequency characteristic is
relatively mild, approximation methods can be used.
The proposed method takes the wave picture as a
starting point rather than currents and voltages.
Distortions can be described by the interference
of several waves arriving by different ways into the

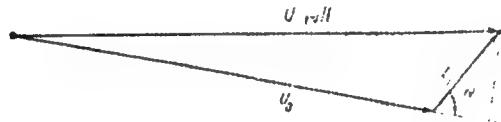
Card 1/9

Phase and Frequency Distortions in Mixers

7772
357/103-5-2-5/26

output field of the system. In the ideal ω - ν - f system only one path exists, but in the real system there may be parasitic paths caused by (a) detuning in the wave guide, (b) double conversion at mirror-frequency or due to harmonics, (c) "squeezing" of the signal from the oscillator to the receiver due to poor shielding. In all these cases an algebraic analysis of the frequency characteristics amounts to a vector analysis of the diagram. The full field at the output of the system is a vector sum (see Fig. 1.)

$$\vec{U}_{full} = \vec{U}_0 + \vec{U}_1 = U_0 \left(1 + \frac{U_1}{U_0} \right). \quad (1)$$



Card 2/9

Fig. 1.

MANUFACTURE OF POLY(1,4-PHENYLENE TEREPHTHALIC ANHYDRIDE)

340 *Journal of Health Politics*

γ is the angle between the two vectors.

$$\frac{\Delta A(\theta)}{\Delta r} = 8.69 \text{ p.c.} \sin \theta$$

APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

For example, if $\mu = 0$ then $\hat{f}^{\mu} = f$

and if $\mu = 1$ then $\hat{f}^{\mu} = f^{\mu}$

$\hat{f}^{\mu} = \mu f + (1-\mu) f^{\mu}$

(4)

and if $\mu = 2$ then $\hat{f}^{\mu} = f^2$

$\hat{f}^{\mu} = \mu f + (1-\mu) f^2$

(5)

and if $\mu = 3$ then $\hat{f}^{\mu} = f^3$

etc.

$$\begin{aligned} \int \Delta \psi \, d\omega_1 &= 0 \\ \int \Delta \psi^2 \, d\omega_1 &= \frac{1}{2} \int \Delta \psi \, d\omega_1 \Delta \psi \, d\omega_1 = 0 \quad (7) \\ \int \Delta \psi^2 \, d\omega_1 &= \frac{1}{2} \int \Delta \psi \, d\omega_1 \Delta \psi \, d\omega_1 = 0 \end{aligned}$$

But this is not the case. For the first term, we have $\int \Delta \psi \, d\omega_1 \neq 0$.
For small values of $\Delta \psi$, we have $\int \Delta \psi \, d\omega_1 \Delta \psi \, d\omega_1 \neq 0$.
Small and large values of $\Delta \psi$ are both possible.

Introduction to Discretization (1/3)

1970-1971

• 100 •

THE JOURNAL OF CLIMATE

$$ds^2 = -c^2 \sin^2\theta \left[\left(\frac{r^2}{a^2} \right)^{1/2} dt^2 - dr^2 \right] + \left(\frac{r^2}{a^2} \right)^{1/2} \left[\left(\frac{r^2}{a^2} \right)^{1/2} d\theta^2 + d\phi^2 \right] + \left(\frac{r^2}{a^2} \right)^{1/2} d\psi^2$$

Analysts, α -Tocopherol, and β -Carotene in the Human Eye

Carries on the work of the *Journal of the American Academy of Religion*.

APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

For the present, the first term in (1) may be

approximated by the first term in (2) for the purpose of representing the initial state.

$$\text{For } \frac{t_0 - t_0^*}{\tau_{\text{int}}} \ll 1, \quad \text{we have}$$

(i) Since the amplitude is small, $\theta_{\text{int}}^*(t_0)$ is small and the first term in (2) is dominant. The first term in (2) is proportional to $\theta_{\text{int}}^*(t_0)$ and the probability is proportional to $\theta_{\text{int}}^*(t_0)^2$. The first term in (2) is proportional to $\theta_{\text{int}}^*(t_0)$ and the probability is proportional to $\theta_{\text{int}}^*(t_0)^2$. The first term in (2) can be written as

$$S_{\text{int}} = \frac{e^2 \hbar}{4 m} \left[\theta_{\text{int}}^* \sin^2 \left(\frac{\theta}{2} \right) + \theta_{\text{int}}^* \cos^2 \left(\frac{\theta}{2} \right) \right] \quad (3)$$

where θ_{int}^* is the initial value of the angle of deflection and $\theta = \theta_{\text{int}}^* + \Delta\theta$. The angle $\Delta\theta$ is small, $\Delta\theta \ll \theta_{\text{int}}^*$, and $\Delta\theta = \Delta T_{\text{int}}/\omega$.

Phase-frequency distribution in MHD

below. In addition, it is the MHD approximation below. In addition, it is the MHD approximation

$$z_{\text{in}} = \left(\frac{1}{4} \sin \theta \right) e^{-i\theta} \left[\lambda \theta \right] \quad (12)$$

for small θ . The term $\lambda \theta$ is the MHD approximation to the effect of the magnetic field on the particle motion.

$$\frac{d\theta}{d\theta} \approx 0, \frac{d\theta}{d\phi} = 2Q \frac{1}{\eta_0 \theta}, \quad (13)$$

It may be anticipated that the effect of the magnetic field on the

$$\dot{\theta}_{\text{in}} = \frac{1}{4} \sin \theta \left(2Q \right) \frac{\lambda \theta - \frac{1}{2} \lambda \theta}{\eta_0 \theta} \quad (14)$$

where

Printed Frequency: 101.00000 MHz

101.00000

101.00000

That disturbance due to passing of the intermediate frequency through a variable capacitor in the trans when due to incident power of 1000 watts. To avoid disturbance at the small signal power, the frequency converter must be able to handle 1000 watts with a minimum of power loss. The power of 1000 watts is permissible. If a power of 1000 watts is placed after a frequency converter, we may say it is of no importance, the use of filter and the active rectifier is recommended. In the case that the author reiterated the importance of the filter and the rectifier between the mixer and the preamplifier, the use of active rectifiers in the case of powerful mixer. There are 4 types of rectifiers.

SUBMITTED: February 14, 1964

Card 1/1

9.3740

7705
307/170-545-1, 1, 1

AUTHORS: Gertsenstern, M. E., Kinder, D. E.

TITLE: Phase Selection in the Diffusion Process of Small
Crystallites

PERIODICAL: Radiotekhnika i elektronika, Moscow, Vol. 30, No. 10, 1985
p. 2200 (USSR)

ABSTRACT: In contrast to conventional crystallization, the process of crystallization in a polydisperse system of small crystallites is characterized by a phase selection mechanism. The research has shown that a polydisperse system differs with respect to the phase of small crystallites selected. Let this be called the "phase selection". The small crystallites with given selected size of a particular crystallite within one degree of freedom (within reference to an unfixed crystal). The process of small crystallite selection is controlled by a parameter, which differs from the usual one, the size of the primary crystallite. The size of the primary crystallite is determined by the size of the small crystallites, which are selected by the phase selection mechanism.

and 1977, and is one of the most frequently used (e.g., L. L. Miedel, Linton, and H. D. Tull, 1977; L. L. Miedel et al., 1977; L. L. Miedel, 1978), and is also used in the present study (see Fig. 1). The present study is the first to use the Linton and Tull (1977) model to predict the development and differentiation of the *Leucaspis* complex in the field. The results of this study will be compared with those of the previous studies to examine the development and differentiation of the *Leucaspis* complex in the field.

$$\ddot{\theta} + 2\dot{\theta}\dot{\phi} + \omega_0^2(1 + q \sin \theta) \sin \theta \dot{\phi}^2 = \omega_0^2 t \cos \Omega t, \quad (1)$$

where $\mu(\cdot)$ is a function of ω and θ and θ is the parameter of the system ($\theta \in \mathbb{R}^n$); $\mu(\cdot)$ is called the frequency response function of the system. The frequency response function $\mu(\cdot)$ is called the transfer function of the system. The frequency response function $\mu(\cdot)$ is called the transfer function of the system.

G. S. H. HUANG

Plane of the Earth's orbit

Plane of the ecliptic

1. $\psi(\tau) = \psi_0 e^{-i\omega_0 \tau}$

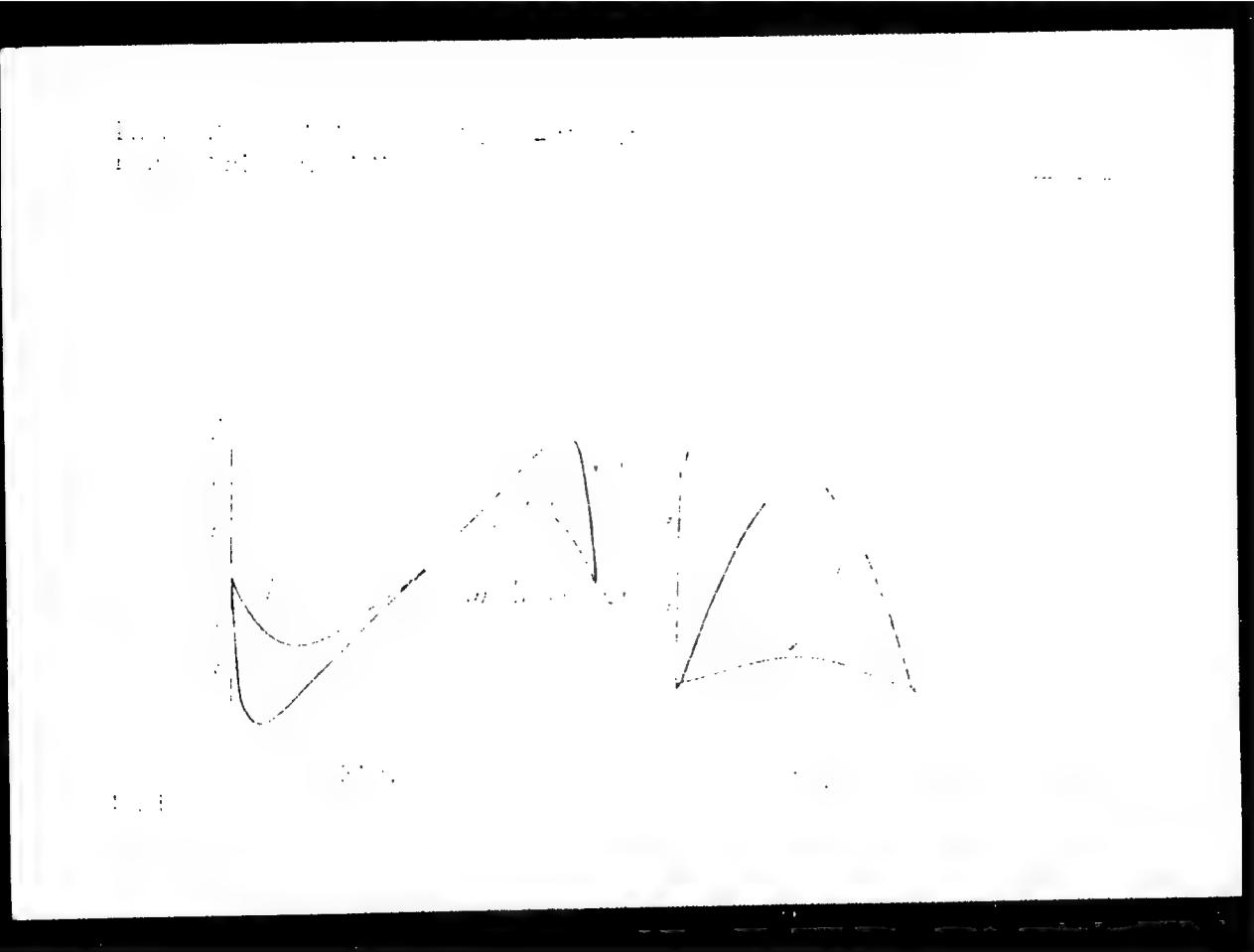
2. $\psi(\tau) = \psi_0 e^{i\omega_0 \tau}$

3. $\psi(\tau) = \psi_0 e^{i\omega_0 \tau} + \psi_1 e^{-i\omega_0 \tau}$

Case 3/1.

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

For τ in \mathcal{H} we have $\varphi(\tau) \in \mathcal{H}$:

$$(\varphi(\tau))^m = \varphi(\tau^m) = \varphi(\tau)$$

For $\tau \in \mathcal{H}$ we have $\varphi(\tau) \in \mathcal{H}$:

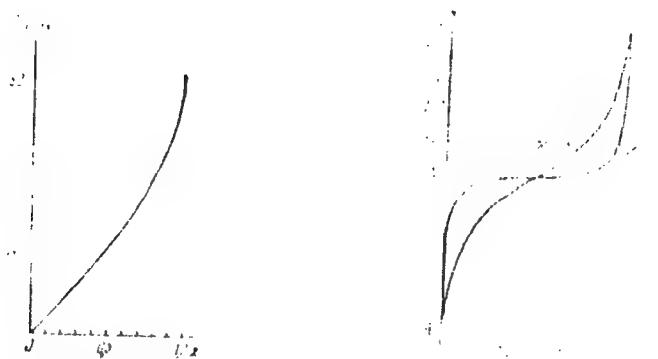
$\varphi(\pi(\tau)) = \pi(\varphi(\tau))$
 $\varphi(\varphi(\tau)) = \varphi(\tau)$
 $\varphi(\tau) = \tau$

$$b = \int_0^1 b(t) dt = \int_0^1 \varphi(b(t)) dt = \int_0^1 b(t) dt = b$$

QED

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

"APPROVED FOR RELEASE: 09/24/2001

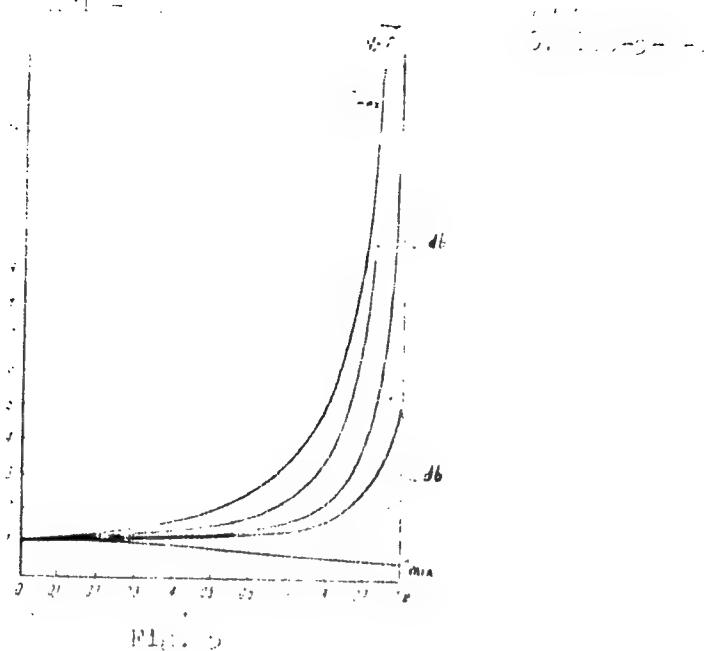
CIA-RDP86-00513R000514920017-2

1. $\mathcal{K} \rightarrow \pi \rightarrow \varphi$
2. $\pi \rightarrow \varphi$

$\mathcal{K} \rightarrow \pi \rightarrow \varphi$
 $\pi \rightarrow \varphi$

APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"



Card 9/17

Printed 10/1/86 by: Gindle-Circuit
Prepared by: Ampt'Filter

1765
200,000,000

and the term with negative frequency, $e^{-i\omega t}$, is excluded. Assuming that only resonance frequencies are significant, put the $\Omega \approx \omega_0$ and let $\nu = \Omega - \nu$ $= -\mu$, for $\nu \approx \omega_0$; $|\mu| \approx \omega_0$ and $\nu \ll \omega_0$. Then the weighting function, ρ , is given by the equation for Ω and $\Omega - \nu$:

$$\rho = e^{i\mu t} + e^{-i\mu t}$$

where

$$\mu = \nu - \Omega \approx -\nu$$

α and β are the constants of the Ω and ν axes, respectively, as shown in Figure 1. Furthermore, Ω and $\Omega - \nu$ must be even functions of μ and ρ must be the product thereof (i.e., $\rho = \Omega \cdot \Omega - \nu$). This is the weight function:

Card 10/1/86

Diagram of the
Polarized Vector

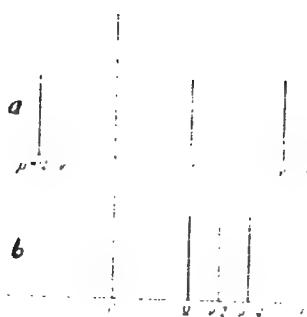


Fig. A

Card 11 (1)

Report of the Committee on Intelligence
House of Representatives

Report of the Committee on Intelligence
House of Representatives
Hearings on the Foreign Intelligence
Surveillance Act of 1978

$$\text{Re}(\eta) = \frac{1}{2}(\eta - \bar{\eta}) = \frac{1}{2}(\eta - \frac{1}{2}(\eta + \bar{\eta}))$$

Im(η) = $\frac{1}{2}(\eta - \bar{\eta})$ is the imaginary part of η and is called the η -part of η .

$$|\eta| = \sqrt{1 - \text{Re}^2(\eta)} = \sqrt{1 - \frac{1}{4}(\eta - \bar{\eta})^2} = \sqrt{1 - \frac{1}{4}(\eta - \bar{\eta})(\eta - \bar{\eta})^*}$$

$$\text{arg} \eta = \text{arg} \frac{\eta - \bar{\eta}}{\sqrt{1 - \text{Re}^2(\eta)}} = \text{arg} \frac{\eta - \bar{\eta}}{|\eta|}$$

Cont'd. 1 of 17

Third, in the case of (3) the frequency spectrum is a stationary one, and the frequency of the oscillations is constant. This is the case of a linear system with a constant load. The load is characterized by a constant value of the parameter μ . The system is considered to be a two-dimensional system, and the two axes are the first one of the load and the frequency. This applies to the amplifier with the dispense of feedback, where it is possible to eliminate beats by separating frequency μ and Ω in the spectral spectrum. In the circuit the oscillations of the frequency Ω with the Ω and $\Omega - \nu = -\mu$, which are mirror images with respect to $\nu/2$, and beats of these frequencies are observed in load resistivity. (4) Amplification of modulated signals by a parametric amplifier. Above, the amplification of a harmonic signal was analyzed. For analysis of a differential signal, the signal will be expanded by a sum: $k = k_1(\omega) + k_2(\omega)\omega$.

Card 13/17

$$k = k_1(\omega) + k_2(\omega)\omega. \quad (3)$$

Phase coherentality of classical propagator
Parameter: $\Delta \mu \ll \mu$

Wavelength of propagator k_1 and k_2 :

$$k_1 = \frac{\omega_0^2}{2D_{11}} \{ \omega_0^2 - \mu^2 - 2\gamma \mu \} \quad (19)$$

$$k_2 = \frac{\omega_0^2 \mu}{D_{11} \gamma} \quad (20)$$

Assuming the amplitudation of Ω is $\Omega = \Omega_0 e^{i\Omega t}$ and taking the expectation value of the propagator $\langle \psi_1 | \psi_2 \rangle$ we find, if Ω_0 is prepared by a coherent displacement $\delta \psi_0 = \delta \psi_0 e^{i\Omega_0 t}$:

$$\begin{aligned} \langle \psi_1 | \psi_2 \rangle &= \int_0^{\infty} e^{i\Omega t} \langle 1 - \delta \psi_0 | \psi_1 \rangle \langle \psi_2 | 1 - \delta \psi_0 \rangle d\omega = \langle \psi_1 | \psi_2 \rangle \\ &= 2e^{i\Omega t} \sqrt{\int_0^{\infty} e^{-i\Omega t} \varphi(\omega - \Omega) d\omega} = \langle \psi_1 | \psi_2 \rangle, \\ &= e^{i\Omega t} \int_0^{\infty} e^{-i\Omega t} \varphi(\omega) d\omega = \langle f(t) - \varphi(t) | f(t) \rangle, \end{aligned}$$

where $f(t) = e^{i\Omega t} \varphi(t)$

Phase Selectivity of Single-Circuit
Parametric Amplifier756
0111-11-17-17-000

$$r_1(t) = \int_{-\infty}^{t_0} e^{i\omega t} \psi(\omega) d\omega$$

Thus, in the case of a signal of any shape, phase selectivity is also present. (6) Noise amplification. White spectrum noise is the totality of incoherent sinusoids with arbitrary phases; their amplification coefficient is (31). With a quadratic detector, it is:

$$|R^2| = \frac{1+x^2}{1-x^2} \quad (29)$$

Consequently, phase selectivity does not play any role in noise amplification. Equation (29) is also valid if not only the phase of the amplified signal, but also the phase of the pumping field is arbitrary (or at

Card 15/17

Phase Selectivity of Single-Circuit
Parametric Amplifier

707/1 1-5-3-10/6

random). It seems that the field of an incoherent source can be used as the pumping field. Noises and distortions of such an amplifier, of course, should be investigated separately. (7) Influence of Phase Selectivity. From the above, it follows that phase selectivity leads to amplitude and phase modulation of the signal being amplified. Pulses at the output of a parametric amplifier are amplitude-modulated. This modulation can be removed with the help of a system of automatic amplitude regulation in the receiver. Analyzing FM of the signal, the spectral method is recommended. Conclusions: (1) A parametric amplifier with one degree of freedom, when amplifying a signal with frequency Ω , causes a beat modulation of the amplified signal, resulting in phase oscillations $\nu - 2\Omega$. (2) Solutions for near-resonance area by simplified equations and complex amplitude methods are identical, and the method of complex amplitudes can be used for the solution of more complicated problems. (3) A parametric amplifier with one degree of freedom is phase-selective, as its instant

Card 16/17

Phase 3: Analysis and Reporting

SUBMITTED: May 1, 1968

GOLD

GERTSENSHTEYN, M. Ye.; VASIL'YEV, V.B.

In regards to S. I. Al'ber and V. I. Bespalov's letter "Diffusion
equation for a statistically nonhomogenous wave guide. Radio-
tekhn. i elektron. 6 no.3:449-450 Mr '61. (MIRA 14:3)
(Wave guides)
(Al'ber, S. I.) (Bespalov, V. I.)

89206

24.4400

S/056/61/040/C01/012/037
B1C2/B204

AUTHOR: Gertsenshteyn, M. Ye.

TITLE: The laws of conservation in the general relativity

PERIODICAL: Zhurnal eksperimental'noy i teoreticheskoy fiziki, v. 40,
no. 1, 1961, 114-122

TEXT: The author deals with two points in the theory of the laws of conservation which are, seen from the mathematical viewpoint not clear: 1) The energy momentum vector $P_i = \int_i^k ds_k$ is in this integral representation not satisfactory, because the vector addition is not defined. 2) In the representation of the coordinate transformation (2):

$\delta x^i = \xi^i(x) = x_j^i(x) \delta \omega^j$, where $\delta \omega^j$ are the parameters of an element of the irreducible group of coordinate transformation (translation or rotation), it is not definitely said what functions $\xi^i(x)$ correspond to the translation. Integrals like the one abovementioned occur in the general relativity when the laws of conservation are being studied. If t_i^k is an energy momen-

Card 1/5

89206

The laws of conservation ...

S/056/61/040/001/012/037
B102/B204

tum pseudotensor, and if composition (integration) is carried out according to components (the coordinates are Euclidean at infinity), then the integral is independent of coordinate system; if, at $x^a \rightarrow \infty$, ξ^i tend \rightarrow const, the integral quantities, which were obtained in the integration of various energy-momentum tensors (which are produced by (2)), coincide. Such a situation, where the mathematical operation employed is not defined, and obtains sense only by the nature of the expression under the integral, is considered to be unsatisfactory by the author. Definition of the integral and the translation is purely geometric, and ought to be independent of the physical content of the problem. For determination of this integral in Riemann geometry, a so-called "free" vector field is introduced, which uniquely (i.e., independent of path) describing the shift of the origin of the coordinates is introduced: $P_i(x) = \hat{C}P_i(x_0)$, where \hat{C} is the operator of the "harmonic" shift. $\xi^i(x)$ is considered to be a vector field, which obeys the following conditions: $\xi^i(x, x_0)$ is a unique function, $\xi^i(x)$ are vectors which are parallel in Euclidean space. Thus it is possible, like above, to put $\xi^i(x) = \hat{C}\xi^i(x_0)$. The harmonic shift is defined in all spaces ✓

Card 2/5

09400

The laws of conservation ...

S/056/61/040/001/012/037
B102/B204

that are topologically equivalent to Euclidean space; however, it differs from a parallel shift. For the "free" vector in a curvilinear pseudoeuclidean space $\nabla_k p^s(x) = 0$ holds, ∇_k denotes a covariant derivative, if k and s are independent, this equation contains 16 conditions. With the definition of the invariant $\zeta = \nabla_k p^k = \text{div}p$, and separation of the symmetric and anti-metric part, $\xi_{ik} = \nabla_k p_i + \nabla_i p_k$, $\eta_{ik} = \nabla_k p_i - \nabla_i p_k$, it is possible to impose onto the vector field $p_i(x)$ the condition $\xi_{ik} = 0$ (which in itself comprises 10 conditions). These conditions have already been studied by V. A. Fok. They are satisfied only in a space of constant curvature ($\nabla_s R = 0$). The conditions (13): $\zeta = 0$, $\eta_{ik} = 0$ (7 conditions) are, on the other hand, satisfied in the case of arbitrary R_{iks}^m . The solution of (13) is given with $p_k = \nabla_k \psi = \partial \psi / \partial x^k$, $\square \psi = 0$. The general-covariant linear differential equations (13) define the geometric operation of a "harmonic" translation

Card 3/5

89026

S/056/61/040/001/012/037
B1C2/B204

The laws of conservation ...

of the vector in a unique manner. There now exists, also in the general case of a space of arbitrary curvature, a preferred system of coordinates, in which the components of the vector remain unchanged in the case of a shift. The condition

$\partial(\sqrt{-g} g^{ik})/\partial x^i = 0; g^{im} \Gamma^k_{im} = 0$ determines the class of the

"harmonic" (preferred) system of coordinates. In such a system, the covariant vector components in harmonic translation do not change, and it is therefore possible to integrate the vectors by the components. Energy-momentum vector, - pseudotensor, energy density, and the Hamiltonian of the system should, therefore, be calculated in such a harmonic system. The case of infinitely small coordinate transformations is studied and the formula hereby for the energy-momentum tensor is applied to the gravitational field. For the canonic energy-momentum tensor, a unique expression is obtained which after symmetrization goes over into the Landau-Lifshits tensor. In conclusion, the case is studied in which the gravitational field may be considered to be a slight perturbation, and the results of the calculations are compared in the various systems of coordinates. The

Card 4/5

59206

The laws of conservation ...

S/056/61/C40/001/012/037
B102/B204

author finally thanks V. L. Bonch-Bruyevich, Professor A. Z. Petrov, A. A. Fedorov, and L. G. Solovey for discussions. There are 10 references: 4 Soviet-bloc and 4 non-Soviet-bloc.

SUBMITTED: October 8, 1959 (initially) and March 9, 1960 (after revision)

X

Card 5/5

26412
 S/056/61/341/001/007/021
 B102/B214

9.9867

AUTHOR: Gertsenshteyn, N. Ye.

TITLE: Wave resonance of light and gravitational waves

PERIODICAL: Zhurnal eksperimental'noy i teoreticheskoy fiziki, v. 41,
 no. 1(7), 1961, 113-114

TEXT: This paper gives an estimate of the energy of gravitational waves produced during the propagation of light in a constant electric or magnetic field. According to general relativity light and gravitational waves propagate with equal velocity, and the corresponding rays coincide with the zero geodesics. That means that, if there exists a linear relationship between light and gravitation waves, wave resonance known in radio physics must appear so that even in weak coupling a significant energy transfer may take place. In the presence of an electromagnetic field a weak gravitational field is described by

$$\square \psi^k = -16\pi c^{-4} \tau^k, \quad \tau^k = 0, \quad \tau^k_{,k} = 0, \quad (1)$$

$$\tau^k = \frac{1}{4\pi} [F^{kl} F_{kl} - \frac{1}{4} \delta^k_l (F^{mn} F_{mn})], \quad \psi^k = h^k_l - \frac{1}{4} h \delta^k_l.$$

Card 1/5

26412
 S/056/61/041/001/007/021
 B102/B214

Wave resonance of light and ...

where τ^{ik} is the energy - momentum tensor of the electromagnetic field, F^{ik} is the electromagnetic field tensor, γ the gravitational constant, and h_{ik} the perturbation of the metric tensor. Eq. (1) is used for investigating the propagation of light (F^{ik} field) in the presence of a strong magnetizing field $F^{(0)ik}$ constant in space and time. The energy - momentum tensor becomes the sum of three terms: square of a constant term, square of the light wave field, and an interference term describing the wave resonance. On neglecting the non-resonance term one obtains the relation

$$\square \psi' = -\frac{8\gamma}{c^2} [F^{(0)kl} F_{kl} - \frac{1}{4} \delta_{ik} (F^{(0)lm} F_{lm})]. \quad (2)$$

If the y -axis is taken in the direction of the wave vector and the wave amplitude is expressed in the units of energy density, one obtains

$$F_{kl} = b(x) \delta_{kl} e^{ikx}, \quad F^{kl} \delta_{kl} = 1, \quad k = \omega/c, \quad (3)$$

$$\psi'^k = a(x) \sqrt{16\pi\gamma/c^4 k^3} \zeta^k e^{ikx}, \quad \zeta_{kl} \zeta^{lk} = 1, \quad \zeta^0 = 0;$$

Card 2/5

26412

S/056/61/041/001/007/021

B102/3214

Wave resonance of light and ...

where the amplitudes f_{kl} and f_{ik} are dimensionless. With this one obtains in the approximation of slowly varying amplitudes: $i \partial a(x) / \partial x$ $= (\gamma/\pi c^4) F^{(0)il} f_{kl} f_{ik} b(x)$. The solution of this equation has the form $a(x) = i \sqrt{\gamma/\pi c^4} f_{kl} \int_i^k \int_0^x F^{(0)il}(s) \cdot b(s) ds + a(0)$, where the integration is made along the ray. If $a(0) = 0$ the external field is constant and the absorption or scattering of the light along the ray is small in the domain considered; i.e. $b(s) = \text{constant}$ so that $|a(x)/b(0)|^2 = (\gamma/\pi c^4) F^{(0)2} T^2$, where T is the time in which the ray traverses the constant field. The amplitude packet was here set equal to one. If the $F^{(0)}$ field is turbulent and random, it can be assumed for the purpose of estimating the energy of the gravitational wave that $F^{(0)}$ is constant along a path of length R_0 (R_0 - correlation radius of the $F^{(0)}$ field) and then changes by jumps and at random. The light amplitude $b(x)$ is practically constant along the ray; the amplitude of the gravitational wave is given by

Card 3/5

~~XX~~

26412
S/056/61/041/001/007/021
B102/E214

Wave resonance of light and ...

$$a(x) = \sum a_n; \quad a_n = i \sqrt{\gamma/\pi c^4} f_M \zeta_n^k \int_{s_{n-1}}^{s_n} F^{(0)II}(s) b(s) ds.$$

The gravitational waves excited at each portion of the path become incoherent. One obtains: $|a(x)/b|^2 = (\gamma/\pi c^3) F^{(0)2} R_0 T$ (7). For interstellar fields one obtains, for example, $|a/b|^2 \sim 10^{-17}$, ($T^{(0)} = 10^{-5} G$, $R_0 = 10$ light years, $T = 10^7$ years). The frequency of the excited gravitational wave is determined by the light frequency. Strong magnetic fields exist also inside the stars, and therefore gravitational waves can be produced. Here the correlation radius $a(x)$ is essentially determined by the free path of the radiation. For the calculation of the intensity of this wave (7) can also be used, but then T is the diffusion time of the energy of the ray in the star transparent to the radiation. It can be shown that (7) represents the ratio of the gravitational and light radiations of the star. Naturally, the intensity of the gravitational radiation is small and is unimportant for the energy balance of the star. There are 3 Soviet-bloc references.

Card 4/5

34044
S/109/62/007/001/025/027
D266/D301

9.3240 (1040, 1139, 1154)

AUTHOR: Rabinovich-Vizel', A. A., and Gertsenshteyn, M. Ye.

TITLE: On the bandwidth of frequency multipliers employing
non-linear capacitance

PERIODICAL: Radiotekhnika i elektronika, v. 7, no. 1, 1962.
175 - 177

TEXT: The purpose of the paper is to determine the bandwidth of frequency multipliers using non-linear elements. The authors first survey available literature and conclude that the efficiency of this type of frequency multiplier has received much attention, but hardly anything has been written on the attainable bandwidth. Next they quote K.M. Johnson's formulas, slightly rearrange them and find for the product of relative bandwidth and optimum efficiency

$$\eta_{\text{opt}} \frac{\Delta f}{f} = \sqrt{b_n^2 + (\omega_1 \tau)^2} \cdot \omega_1 \tau. \quad (6)$$

where b_n depends on the nonlinear characteristics of the diode em-

Card 1/2

424
S/109/62/CC7/001/026/027
D266/D301

On the bandwidth of frequency ...

ployed. ω_1 - fundamental frequency, τ - time constant of the diode, n - factor of multiplication. For a lossless diode

$$\tau = 0, \eta = 1, \frac{\Delta f}{f} = b_n, (Q_{D1} b_n)^2 \gg 1 \quad (7)$$

where Q_{D1} - quality of the diode at the frequency ω_1 . In this case the bandwidth is dependent on n . If the losses are large $Q_{D1} b_n \ll 1$, the bandwidth is mainly determined by the losses and independent of the harmonic number. If non-linear resistances are used there is no difficulty with bandwidth because broadband matching is possible. There are 5 references: 1 Soviet-bloc and 4 non-Soviet-bloc. The 4 most recent references to the English-language publications read as follows: C.H. Page, J. Res. Nat. Bur. Standards, 1956, 56, 4, 179; G. Luetgenau, M.V. Duffin, and P.H. Dirnbach, IRE Wescon Convention Record, 1960, part 3, 13; P.M. Fitzgerald, T.H. Lee, M.S. Moy, E.C. Powers and J.J. Younger, IRE Wescon Convention Record, 1960, part 2, 43; K.M. Johnson, IRE Trans., 1960, MTT-8, 5, 525.

SUBMITTED: July 20, 1961

Card 2/2

2/100/49/007/003/005/029
5234/0202

AUTHORS: Gertsenshtern, N.Ye., and Kinkin, B.Ye

TITLE: Stability of the super-regenerative regime of an amplifier with complex networks

PERIODICAL: Radiotekhnika i elektronika, v. 7, no. 5, 1962,
397 - 403

TEXT: The authors formulate equations for a parametric amplifier with variable capacity without frequency transformation, considering it as an n-terminal network. For the case of a two-circuit non-degenerate regenerative amplifier, an equation of Hill's type is deduced from the general equations; the stability of the solutions is determined by that of the solutions of the corresponding homogeneous equation. It is found that if a complicated input filter is used, whose band is not much wider than that of the amplifier, the domains of stability depend essentially on the parameters of super-regeneration. The case of an input filter consisting of two equal links is considered as an example; the homogeneous equation is reduced

Card 1/2